



Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

General Description

The MAX98314 mono 3.2W Class D amplifier provides Class AB audio performance with Class D efficiency. This device offers five selectable gain settings (0dB, 3dB, 6dB, 9dB, and 12dB) set by a single gain-select input (GAIN).

Active emissions limiting (AEL) edge rate and overshoot control circuitry and a filterless spread-spectrum modulation (SSM) scheme greatly reduce EMI and eliminate the need for output filtering found in traditional Class D devices.

The IC's low 0.95mA at 3.7V, 1.2mA at 5.0V quiescent current extends battery life in portable applications.

Highly linear, integrated input coupling capacitors (C_{IN}) reduce solution size and provide excellent THD+N, PSRR, and CMRR performance at low frequencies vs. standard Class D amplifiers using external input capacitors.

The IC is available in a small 9-bump, 0.3mm pitch WLP (1.0mm x 1.0mm x 0.80mm) package and is specified over the -40° C to $+85^{\circ}$ C extended temperature range.

Applications

Features

- - ♦ f_C = 200Hz (12dB)
- Delivers High Output Power at 10% THD+N
 - \Rightarrow 3.2W into 4Ω, V_{PVDD} = 5V
 - ♦ 960mW into 8Ω, V_{PVDD} = 3.7V
- Ultra-Low Noise: 19µV
- Eliminates Output Filtering Requirement
 Spread Spectrum and Active Emissions Limiting
- Click-and-Pop Suppression
- Thermal and Overcurrent Protection
- Low Current Shutdown Mode
- Small, Space-Saving Package

Simplified Block Diagram



Ordering Information appears at end of data sheet.



For related parts and recommended products to use with this part, refer to: www.maxim-ic.com/MAX98314.related

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For pricing, delivery, and ordering information, please contact Maxim Direct at 1-888-629-4642, or visit Maxim's website at www.maxim-ic.com.

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2.5V TO 5.5V ____ 0.1µF 10µF* PVDD A3 SHDN C1 UVL0/POWER CLICK-AND-POP MANAGEMENT SUPPRESSION ///XI// PVDD MAX98314 OUT-A1 GAIN B3 LOW-EMI DRIVER C3 IN+ PGND = CLASS D MODULATOR PVDD C2 IN-0UT+ A2 LOW-EMI DRIVER PGND ± B1 PGND <u>+</u> *SYSTEM BULK CAPACITANCE.

Functional Diagram/Typical Application Circuit

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ABSOLUTE MAXIMUM RATINGS

PVDD, IN+, IN-, SHDN, GAIN to	• PGND 0.3V to +6V
OUT+, OUT- to PGND	0.3V to (V _{PVDD} + 0.3V)
Continuous Current In/Out of PN	/DD, PGND, OUT750mA
Continuous Input Current (all ot	her pins)±20mA
Duration of Short Circuit Betwee	en
OUT_ to PVDD, PGND	Continuous
Between OUT+ and OUT- Pir	nsContinuous

Continuous Power Dissipation ($T_A = +70^{\circ}C$) f	or Multilayer Board
WLP (derate 10.64mW/°C above +70°C)	851mW
Junction Temperature	+150°C
Operating Temperature Range	40°C to +85°C
Storage Temperature Range	65°C to +150°C
Soldering Temperature (reflow)	+260°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

PACKAGE THERMAL CHARACTERISTICS (Note 1)

WLP

Note 1: Package thermal resistances were obtained using the method described in JEDEC specification JESD51-7, using a fourlayer board. For detailed information on package thermal considerations, refer to <u>www.maxim-ic.com/thermal-tutorial</u>.

ELECTRICAL CHARACTERISTICS

 $(V_{PVDD} = V_{SHDN} = V_{GAIN} = 5V, V_{PGND} = 0V, A_V = 6dB (GAIN = PVDD), R_L = \infty, R_L connected between OUT+ to OUT-, AC measurement bandwidth 20Hz to 22kHz, T_A = T_{MIN} to T_{MAX}$, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2, 3)

PARAMETER	SYMBOL		CONDITIO	ONS	MIN	ТҮР	MAX	UNITS	
AMPLIFIER CHARACTERISTIC	S								
Supply Voltage Range	V _{PVDD}	Guarantee	Guaranteed by PSRR test		2.5		5.5	V	
Undervoltage Lockout	UVLO	PVDD falli	ng			1.8	2.2	V	
Quiescent Current		V _{PVDD} = 5	5V			1.2	1.8	mA	
	IPVDD	$V_{PVDD} = 3$	3.7V			0.95			
Shutdown Supply Current	ISHDN	V _{SHDN} = 0	$V, T_A = +25^{\circ}$	C		< 0.1	10	μA	
Turn-On Time	t _{ON}					3.7	10	ms	
Bias Voltage	V _{BIAS}					V _{PVDD} /2		V	
	Av	f = 1kHz	GAIN conne	cted to PGND	11.75	12	12.25		
			GAIN connected to PGND		8.75	9	9.25	dB	
			through 100k Ω ±5% resistor			0			
Voltage Gain			GAIN connected to PVDD		5.75	6	6.25		
			GAIN connected to PVDD through $100k\Omega \pm 5\%$ resistor		2.75	3	3.25		
			GAIN uncon		-0.25	0	+0.25		
Input Capacitance	C _{IN}	All gains	1			0.011		μF	
				$A_V = 12 dB$		199			
				$A_V = 9 dB$		139		1	
Highpass Corner Frequency	f _C	-3dB dow	n	$A_V = 6 dB$	63	100	189	– Hz	
				$A_V = 3dB$		70		1	
				$A_V = 0 dB$		50		1	



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ELECTRICAL CHARACTERISTICS (continued)

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PARAMETER	SYMBOL	CONDITIO	ONS	MIN	ТҮР	MAX	UNITS
Common-Mode Rejection Ratio	CMRR	f _{IN} = 1kHz, input referred			67		dB
Output Offset Voltage	V _{OS}	$T_{A} = +25^{\circ}C$ (Note 4)			±1	±3	mV
Click-and-Pop Level		$R_L = 8\Omega + 68\mu$ H, peak voltage, T _A = +25°C, A-weighted, 32	Into shutdown		-59		dBV
	K _{CP}	samples per second, $T_A = +25^{\circ}C$ (Notes 4, 5)	Out of shutdown		-82		UBV
		V _{PVDD} = 2.5V to 5.5V, T	A = +25°C	70	90		
Power-Supply Rejection Ratio	PSRR		f = 217Hz		74		dB
(Note 4)		$V_{RIPPLE} = 200 m V_{P-P}$	f = 1kHz		72		
			f = 20kHz		49		
		THD+N = 10%	$V_{PVDD} = 5.0V$		3.2		_
		f = 1kHz	$V_{PVDD} = 4.2V$		2.2		
		$R_{L} = 4\Omega + 33\mu H$	$V_{PVDD} = 3.7V$		1.7		
		THD+N = 1% f = 1kHz R _L = 4 Ω + 33 μ H	$V_{PVDD} = 5.0V$		2.6		
Output Power	Роит		$V_{PVDD} = 4.2V$		1.8]
			$V_{PVDD} = 3.7V$		1.4		1
		THD+N = 10% f = 1kHz R _L = 8 Ω + 68 μ H	$V_{PVDD} = 5.0V$		1.8		W
			$V_{PVDD} = 4.2V$		1.2		1
			$V_{PVDD} = 3.7V$		0.96		
		THD+N = 1% f = 1kHz R _L = 8 Ω + 68 μ H	$V_{PVDD} = 5.0V$		1.4		1
			$V_{PVDD} = 4.2V$		1		
			$V_{PVDD} = 3.7V$		0.8		1
Total Harmonic Distortion Plus		f 11/1 -	$R_{L} = 4\Omega,$ P _{OUT} = 1W		0.03	0.1	0(
Noise	THD+N	f _{IN} = 1kHz	$R_{L} = 8\Omega$ $P_{OUT} = 0.725W$		0.03		~ %
			$A_V = 12 dB$		31		
			$A_V = 9 dB$		26		
Output Noise	V _N	A-weighted (Note 4)	$A_V = 6 dB$		23		μV _{RMS}
			$A_V = 3dB$		21		
			$A_V = 0 dB$		19		1
Efficiency	η	$R_L = 8\Omega$, $P_{OUT} = 1.8W$,	f = 1kHz		93		%
Oscillator Frequency	fosc				300		kHz
Spread-Spectrum Bandwidth					20		kHz
Current Limit					2.8		A
Thermal Shutdown Level	1				155		°C



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ELECTRICAL CHARACTERISTICS (continued)

 $(V_{PVDD} = V_{SHDN} = V_{GAIN} = 5V, V_{PGND} = 0V, A_V = 6dB (GAIN = PVDD), R_L = \infty, R_L connected between OUT+ to OUT-, AC measurement bandwidth 20Hz to 22kHz, T_A = T_{MIN}$ to T_MAX, unless otherwise noted. Typical values are at T_A = +25°C.) (Note 2, 3)

PARAMETER	SYMBOL	CONDITIONS	MIN	ТҮР	MAX	UNITS
Thermal Hysteresis				15		°C
DIGITAL INPUT (SHDN)						
Input Voltage High	V _{INH}	$V_{PVDD} = 2.5V$ to $5.5V$	1.4			V
Input Voltage Low	V _{INL}	$V_{PVDD} = 2.5V$ to 5.5V			0.4	V
Input Leakage Current		$T_A = +25^{\circ}C$			±1	μA

Note 2: All devices are 100% production tested at $T_A = +25^{\circ}C$. Specifications over temperature limits are guaranteed by design. **Note 3:** Testing performed with a resistive load in series with an inductor to simulate an actual speaker load. For $R_L = 4\Omega$,

L = 33 μ H. For R_L = 8 Ω , L = 68 μ H.

Note 4: Amplifier inputs AC-coupled to ground.

Note 5: Mode transitions controlled by SHDN control pin.

Typical Operating Characteristics

 $(V_{PVDD} = V_{SHDN} = 5.0V, V_{PGND} = 0V, A_V = 6dB, R_L = \infty, R_L$ connected between OUT+ to OUT-, AC measurement bandwidth 20Hz to 22kHz, T_A = +25°C, unless otherwise noted.)



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Typical Operating Characteristics (continued)

(V_{PVDD} = V_{SHDN} = 5.0V, V_{PGND} = 0V, A_V = 6dB, R_L = ∞, R_L connected between OUT+ to OUT-, AC measurement bandwidth 20Hz to 22kHz, $T_A = +25^{\circ}C$, unless otherwise noted.)



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Typical Operating Characteristics (continued)

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Typical Operating Characteristics (continued)

FREQUENCY (Hz)

 $(V_{PVDD} = V_{\overline{SHDN}} = 5.0V, V_{PGND} = 0V, A_V = 6dB, R_L = \infty, R_L \text{ connected between OUT+ to OUT-, AC measurement bandwidth 20Hz to 22kHz, T_A = +25°C, unless otherwise noted.)$



FREQUENCY (MHz)

Mono 3.2W Class D Amplifier with Integrated Input Coupling Capacitors

Pin Configuration



Pin Description

BUMP	NAME	FUNCTION
A1	OUT-	Negative Speaker Output
A2	OUT+	Positive Speaker Output
A3	PVDD	Power Supply. Bypass PVDD with a 0.1µF and 10µF capacitor to PGND.
B1	PGND	Power Ground
B2	N.C.	No Connection. Can be left unconnected or connected to PGND.
B3	GAIN	Gain Select. See Table 1 for GAIN settings.
C1	SHDN	Active-Low Shutdown Input. Drive SHDN low to place the device in shutdown.
C2	IN-	Inverting Audio Input
C3	IN+	Noninverting Audio Input

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Detailed Description

The MAX98314 features low quiescent current, a lowpower shutdown mode, comprehensive click-and-pop suppression, and excellent RF immunity.

The IC offers Class AB audio performance with Class D efficiency in a minimal board-space solution. The Class D amplifier features spread-spectrum modulation, edgerate, and overshoot control circuitry that offers significant improvements to switch-mode amplifier radiated emissions.

The amplifier features click-and-pop suppression that reduces audible transients on startup and shutdown. The amplifier additionally includes thermal overload and short-circuit protection.

Highly linear, integrated input coupling capacitors (C_{IN}) reduce solution size and provide excellent THD+N, PSRR, and CMRR performance at low frequencies vs. standard Class D amplifiers using external input capacitors.

Class D Speaker Amplifier

The IC's filterless Class D amplifier offers much higher efficiency than Class AB amplifiers. The high efficiency of a Class D amplifier is due to the switching operation of the output stage transistors. Any power loss associated with the Class D output stage is mostly due to the I²R loss of the MOSFET on-resistance and quiescent switching current overhead.

Ultra-Low EMI Filterless Output Stage

Traditional Class D amplifiers require the use of external LC filters, or shielding, to meet electromagnetic interference (EMI) regulation standards. Maxim's patented active emissions limiting edge-rate control circuitry and spread-spectrum modulation reduces EMI emissions, while maintaining up to 93% efficiency.

The spread-spectrum modulation mode flattens wideband spectral components, while proprietary techniques ensure that the cycle-to-cycle variation of the switching period does not degrade audio reproduction or efficiency. The IC's spread-spectrum modulator randomly varies the switching frequency by ± 20 kHz around the center frequency (300kHz). Above 10MHz, the wideband spectrum looks like noise for EMI purposes (Figure 1).

Amplifier Current Limit

If the output current of the speaker amplifier exceeds the current limit (2.8A typ), the IC disables the outputs for approximately 100µs. At the end of 100µs, the outputs are reenabled. If the fault condition still exists, the IC continues to disable and reenable the outputs until the fault condition is removed.

Selectable Amplifier Gain

The IC offers five programmable gain settings, selectable by a single gain input (GAIN).

Table 1. GAIN Selection

GAIN PIN	MAXIMUM GAIN (dB)
Connect to PGND	12
Connect to PGND through 100k Ω ±5%	9
Connect to PVDD	6
Connect to PVDD through 100k Ω ±5%	3
Unconnected	0

Integrated Input Coupling Capacitors (CIN)

The IC integrates two 0.011 μ F input coupling capacitors, C_{IN}. The input coupling capacitors, in conjunction with the amplifier's internal input resistance (R_{IN}), form a first-order highpass filter that removes the DC bias from the incoming signal. These capacitors allow the amplifier to bias the signal to an optimum DC level.



Figure 1. EMI Performance with 60cm of Speaker Cable, No Output Filter



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Assuming zero source impedance, the -3dB corner frequency, f_{-3dB} , is:

$f_{-3dB} = 1/2\pi R_{IN}C_{IN} [Hz]$

The 100ppm/V voltage coefficient of the integrated input coupling capacitor results in excellent low-frequency THD+N performance. Figure 2 illustrates the superior linearity of the IC's integrated input coupling capacitors compared to a similar amplifier with external 0.01µF X7R and X5R 0402 input coupling capacitors.



Figure 2. Low-Frequency THD+N Performance

Shutdown

The IC features a low-power shutdown mode, drawing < 0.1 μ A (typ) of supply current. Drive SHDN low to put the IC into shutdown.

Click-and-Pop Suppression

The speaker amplifier features Maxim's comprehensive click-and-pop suppression. During startup, the clickand-pop suppression circuitry reduces any audible transient sources internal to the device. When entering shutdown, the differential speaker outputs ramp down to PGND quickly and simultaneously.

Applications Information

Filterless Class D Operation

Traditional Class D amplifiers require an output filter. The filter adds cost and size, and decreases efficiency and THD+N performance. The IC's filterless modulation scheme does not require an output filter.

Because the switching frequency of the IC is well beyond the bandwidth of most speakers, voice coil movement due to the switching frequency is very small. Use a speaker with a series inductance > 10μ H. Typical 8Ω speakers exhibit series inductances in the 20μ H to 100μ H range.

Speaker Amplifier Power-Supply Input (PVDD) PVDD powers the speaker amplifier and ranges from 2.5V to 5.5V. Bypass PVDD with a 0.1μ F and 10μ F capacitor to PGND. Apply additional bulk capacitance at the device if long input traces between PVDD and the power source are used.

Layout and Grounding

Proper layout and grounding are essential for optimum performance. Good grounding improves audio performance and prevents switching noise from coupling into the audio signal.

Use wide, low-resistance output traces. As the load impedance decreases, the current drawn from the device increases. At higher current, the resistance of the output traces decrease the power delivered to the load. For example, if 2W is delivered from the device output to a 4 Ω load through 100m Ω of total speaker trace, 1.904W is delivered to the speaker. If power is delivered through 10m Ω of total speaker trace, 1.99W is delivered to the speaker. Wide output, supply, and ground traces also improve the power dissipation of the device.

The IC is inherently designed for excellent RF immunity. For best performance, add ground fills around all signal traces on top or bottom PCB layers.

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WLP Applications Information

For the latest application details on WLP construction, dimensions, tape carrier information, PCB techniques, bump-pad layout, and recommended reflow temperature profile, as well as the latest information on reliability testing results, refer to <u>Application Note 1891</u>: <u>Wafer-Level Packaging (WLP) and Its Applications</u>. Figure 3 shows the dimensions of the WLP balls used on the IC.

Ordering Information

PART	TEMP RANGE	PIN-PACKAGE
MAX98314EWL+	-40°C to +85°C	9 WLP

+Denotes a lead(Pb)-free/RoHS-compliant package.



Figure 3. WLP Ball Dimensions



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Package Information

For the latest package outline information and land patterns (footprints), go to <u>www.maxim-ic.com/packages</u>. Note that a "+", "#", or "-" in the package code indicates RoHS status only. Package drawings may show a different suffix character, but the drawing pertains to the package regardless of RoHS status.

PACKAGE TYPE	PACKAGE CODE	OUTLINE	NO.		LAND PATTERN N			
9 WLP (0.3mm pitch)	W90A0+1	<u>21-0539</u>	2		Refer Application N		efer to ion Note 189 ⁻	
PIN 1 INDICATOR	MARKING AI		<u> </u>		A A A D D D E I E I E S E S S E S E	0.73 1 0.14 2 0.59 3 0.04 Ø 0.2 0.60 0.60 0.30 0 0.00	0 BASIC 1 ±0.03 0 BASIC 0 BASIC 0 BASIC	
		PKG. CODE	E	E MAX	[MIN	D MAX	DEPOPULAT BUMPS	
. E1 .		W90A0+1	0.95	0.98	0.95	0.98	NONE	
SE	-							
		NOTES: 1. Terminal pitch is c 2. Outer dimension i 3. All dimensions in r 4. Marking shown is 5. Tolerance is ± 0.02 6. All dimensions ap 7. Front - side finish c	nillimeter for pack 2 unless s ply to Pb can be e	r. age orie pecified Free (+) ither Blac	ntation re otherwis package ck or Cle	eferenc e. e codes ar.	e only. 5 only.	
		 Terminal pitch is c Outer dimension i All dimensions in r Marking shown is Tolerance is ± 0.02 	nillimeter for pack 2 unless s ply to Pb can be e	r. age orie pecified Free (+) ither Black Market Package 9 bump	ntation re otherwis package ck or Cle	eferenc e. e codes ar.	e only. s only.	

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Revision History

EVISION NUMBER	REVISION DATE	DESCRIPTION	PAGES CHANGED
0	11/11	Initial release	—

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time. The parametric values (min and max limits) shown in the Electrical Characteristics table are guaranteed. Other parametric values quoted in this data sheet are provided for guidance.

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